

wave equation for \mathbf{H} can be rewritten in the form

$$\frac{\partial^2 \mathbf{H}}{\partial x^2} + \frac{\partial^2 \mathbf{H}}{\partial y^2} + \frac{\partial^2 \mathbf{H}}{\partial z^2} = \gamma^2 \mathbf{H}$$

Of particular interest are solutions (*plane waves*) that depend on only one spatial coordinate, say z . Then the equation becomes

$$\frac{d^2 \mathbf{H}}{dz^2} = \gamma^2 \mathbf{H}$$

which, for an assumed time dependence $e^{j\omega t}$, is the vector analog of the one-dimensional scalar wave equation. Solutions are as above, in terms of the propagation constant γ .

$$\mathbf{H}(z, t) = H_0 e^{\pm \gamma z} e^{j\omega t} \mathbf{a}_H$$

The corresponding solutions for the electric field are

$$\mathbf{E}(z, t) = E_0 e^{\pm \gamma z} e^{j\omega t} \mathbf{a}_E$$

The fixed unit vectors \mathbf{a}_H and \mathbf{a}_E are orthogonal and neither field has a component in the direction of propagation. This being the case, one can rotate the axes to put one of the fields, say \mathbf{E} , along the x axis. Then from Maxwell's equation (2) it follows that \mathbf{H} will lie along the $\pm y$ axis for propagation in the $\pm z$ direction.

EXAMPLE 1. Given the field $\mathbf{E} = E_0 e^{-\gamma z} \mathbf{a}_E$ (time dependence suppressed), show that \mathbf{E} can have no component in the propagation direction, $+\mathbf{a}_z$.

The cartesian components of \mathbf{a}_E are found by projection:

$$\mathbf{E} = E_0 e^{-\gamma z} [(\mathbf{a}_E \cdot \mathbf{a}_x) \mathbf{a}_x + (\mathbf{a}_E \cdot \mathbf{a}_y) \mathbf{a}_y + (\mathbf{a}_E \cdot \mathbf{a}_z) \mathbf{a}_z]$$

From $\nabla \cdot \mathbf{E} = 0$,

$$\frac{\partial}{\partial z} E_0 e^{-\gamma z} (\mathbf{a}_E \cdot \mathbf{a}_z) = 0$$

which can hold only if $\mathbf{a}_E \cdot \mathbf{a}_z = 0$. Consequently, E has no component in \mathbf{a}_z .

The plane wave solutions obtained above depend on the properties μ , ϵ , and σ of the medium, because these properties are involved in the propagation constant γ .

14.4 SOLUTIONS FOR PARTIALLY CONDUCTING MEDIA

For a region in which there is some conductivity but not much (e.g., moist earth, seawater), the solution to the wave equation in \mathbf{E} is taken to be

$$\mathbf{E} = E_0 e^{-\gamma z} \mathbf{a}_x$$

Then, from (2) of Section 14.2,

$$\mathbf{H} = \sqrt{\frac{\sigma + j\omega\epsilon}{j\omega\mu}} E_0 e^{-\gamma z} \mathbf{a}_y$$

The ratio E/H is characteristic of the medium (it is also frequency-dependent). More specifically for waves $\mathbf{E} = E_x \mathbf{a}_x$, $\mathbf{H} = H_y \mathbf{a}_y$ which propagate in the $+z$ direction, the *intrinsic impedance*, η , of the medium is defined by

$$\eta = \frac{E_x}{H_y}$$

$$\eta = \sqrt{\frac{j\omega\mu}{\sigma + j\omega\epsilon}}$$

Thus

where the correct square root may be written in polar form, $|\eta|/\underline{\theta}$, with

$$|\eta| = \frac{\sqrt{\mu/\epsilon}}{\sqrt{1 + \left(\frac{\sigma}{\omega\epsilon}\right)^2}} \quad \tan 2\theta = \frac{\sigma}{\omega\epsilon} \quad \text{and} \quad 0^\circ < \theta < 45^\circ$$

(If the wave propagates in the $-z$ direction, $E_x/H_y = -\eta$. In effect, γ is replaced by $-\gamma$ and the other square root used.)

Inserting the time factor $e^{j\omega t}$ and writing $\gamma = \alpha + j\beta$ results in the following equations for the fields in a partially conducting region:

$$\begin{aligned} \mathbf{E}(z, t) &= E_0 e^{-\alpha z} e^{j(\omega t - \beta z)} \mathbf{a}_x \\ \mathbf{H}(z, t) &= \frac{E_0}{|\eta|} e^{-\alpha z} e^{j(\omega t - \beta z - \theta)} \mathbf{a}_y \end{aligned}$$

The factor $e^{-\alpha z}$ attenuates the magnitudes of both \mathbf{E} and \mathbf{H} as they propagate in the $+z$ direction. The expression for α , (5) of Section 14.2, shows that there will be some attenuation unless the conductivity σ is zero, which would be the case only for perfect dielectrics or free space. Likewise, the phase difference θ between $\mathbf{E}(z, t)$ and $\mathbf{H}(z, t)$ vanishes only when σ is zero.

The velocity of propagation and the wavelength are given by

$$\begin{aligned} u &= \frac{\omega}{\beta} = \frac{1}{\sqrt{\frac{\mu\epsilon}{2} \left(\sqrt{1 + \left(\frac{\sigma}{\omega\epsilon}\right)^2} + 1 \right)}} \\ \lambda &= \frac{2\pi}{\beta} = \frac{2\pi}{\omega \sqrt{\frac{\mu\epsilon}{2} \left(\sqrt{1 + \left(\frac{\sigma}{\omega\epsilon}\right)^2} + 1 \right)}} \end{aligned}$$

If the propagation velocity is known, $\lambda f = u$ may be used to determine the wavelength λ . The term $(\sigma/\omega\epsilon)^2$ has the effect of reducing both the velocity and the wavelength from what they would be in either free space or perfect dielectrics, where $\sigma = 0$. Observe that the medium is *dispersive*: waves with different frequencies ω have different velocities u .

14.5 SOLUTIONS FOR PERFECT DIELECTRICS

For a perfect dielectric, $\sigma = 0$, and so

$$\alpha = 0 \quad \beta = \omega\sqrt{\mu\epsilon} \quad \eta = \sqrt{\frac{\mu}{\epsilon}}/0^\circ$$

Since $\alpha = 0$, there is no attenuation of the \mathbf{E} and \mathbf{H} waves. The zero angle on η results in \mathbf{H} being in time phase with \mathbf{E} at each fixed location. Assuming \mathbf{E} in \mathbf{a}_x and propagation in \mathbf{a}_z , the field equations may be obtained as limits of those in Section 14.4:

$$\begin{aligned} \mathbf{E}(z, t) &= E_0 e^{j(\omega t - \beta z)} \mathbf{a}_x \\ \mathbf{H}(z, t) &= \frac{E_0}{\eta} e^{j(\omega t - \beta z)} \mathbf{a}_y \end{aligned}$$

The velocity and the wavelength are

$$u = \frac{\omega}{\beta} = \frac{1}{\sqrt{\mu\epsilon}} \quad \lambda = \frac{2\pi}{\beta} = \frac{2\pi}{\omega\sqrt{\mu\epsilon}}$$